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Research

New Jersey

mm-wave beam search

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Outline

- Introduction
- System Model
- Beam Search
- Beam Tracking
- Simulation Results
- Discussion
- References

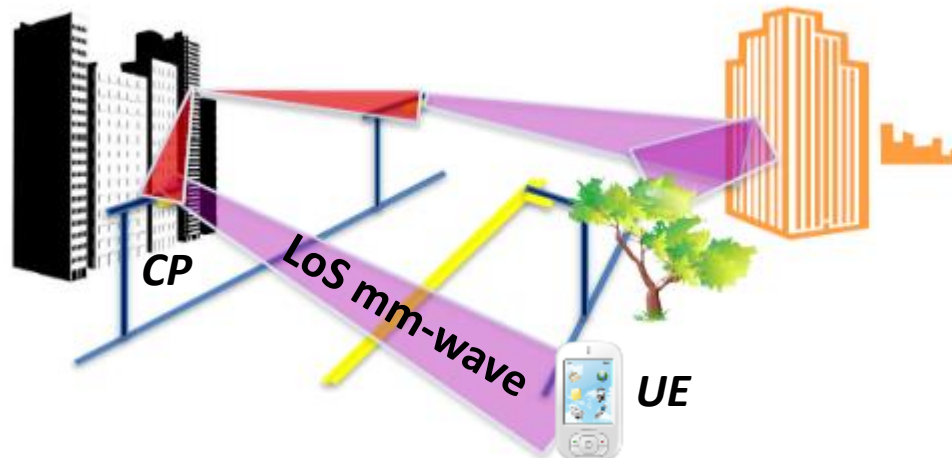
Introduction: mm-wave access

Advantages:

- Large Bandwidths
- High Antenna Density
- Integration with LTE

Challenges:

- High Path Loss
- Need for Beam Search
- Need for Beam Tracking



System Model: MIMO

$$\mathbf{y}[m] = \mathbf{H}[m]\mathbf{x}[m] + \mathbf{w}[m]$$

where $\mathbf{H}[m] \in \mathbb{C}^{N_R \times N_T}$, $\mathbf{w} \sim \mathcal{CN}(0, N_0 \mathbf{I})$

At time m :

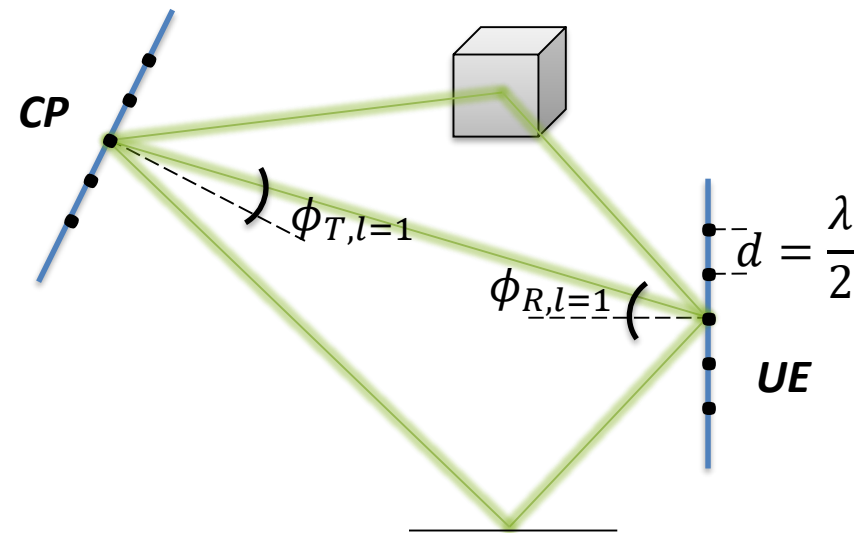
$$\mathbf{H} = \sum_{l=1}^L g_l \mathbf{u}_R[\theta_{R,l}] \mathbf{u}_T^*[\theta_{T,l}]$$

where $g_l \sim \mathcal{CN}(0, \sigma^2)$, $\theta_l = \frac{d}{\lambda} \sin(\phi_l)$

and $\mathbf{u}[\theta] = \begin{bmatrix} e^{-j2\pi\frac{\theta k}{N}} \end{bmatrix}$ with

$$k \in \left\{ i - \frac{N-1}{2} : i = 0, \dots, N-1 \right\}.$$

Uniform Linear Array (ULA)



System Model: Channel

- Channel model is based on outdoor measurements in NYC [2]:

Carrier: 28 GHz **BW:** 1.3 GHz

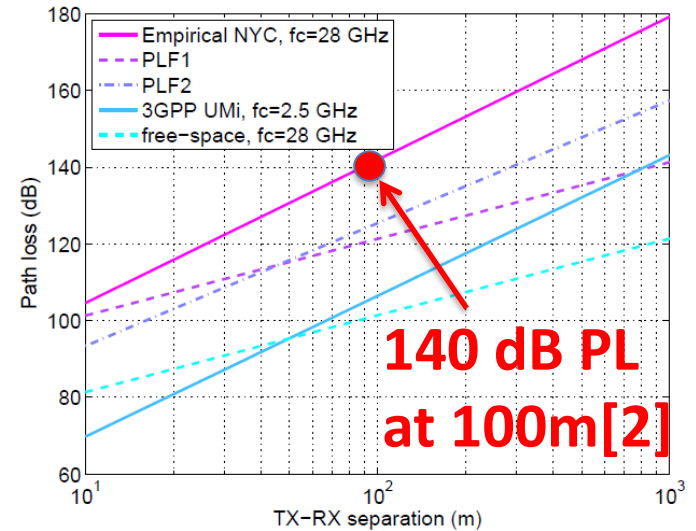
P_{TX} (UE): 20dBm **P_{TX} (CP):** 30dBm

Path Loss: 140dBm@100m

Noise Floor: -83dBm@200MHz

Antenna Gain: $10 \log_{10}(N_T \times N_T)$ (2D array)

T_{COH} = 0.15-2ms **F_{COH}** = 10-100 MHz



$$\mathbf{H} = \sum_{l=1}^L g_l \mathbf{u}_R[\theta_{R,l}] \mathbf{u}_T^*[\theta_{T,l}]$$

$$g_l \sim \mathcal{CN}(0, GK), GK = \frac{K}{L} 10^{-0.1PLK}$$

$$\theta \sim \text{Unif}\left[-\frac{\pi}{2}, +\frac{\pi}{2}\right]$$

| Parameter | Value |
|---------------------|---|
| K (# of clusters) | 3 |
| L (# of subpaths) | 100 |
| PL_K (path loss) | $75.85 + 37.3 \log_{10} d + SK$ |
| S_K (shadowing) | $\sim \text{LogN}(0, \sigma^2), \sigma = 8.36 \text{ dB}$ |

System Model: Beam Search and Tracking

- Beam Search:

$$(\hat{\theta}_T, \hat{\theta}_R) = \underset{\{(\theta_T, \theta_R): -\frac{\pi}{2} \leq \theta_{T,R} \leq \frac{\pi}{2}\}}{\operatorname{argmax}} \mathbb{E} [|\mathbf{w}_R^*(\theta_R, m) \mathbf{H}[m] \mathbf{w}_T(\theta_T, m)|]$$

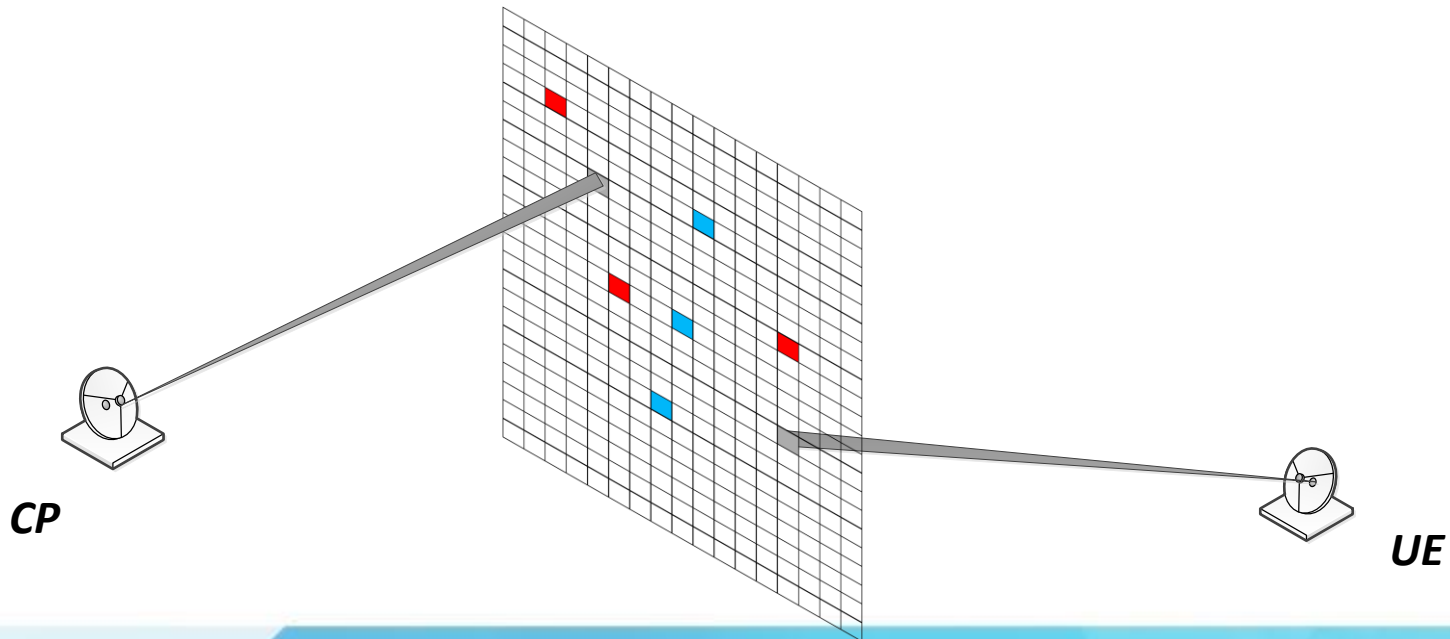
where $\mathbf{w}(\theta, m) = \left[e^{-j2\pi \frac{\theta(m)k}{N}} \right]$ are the beamforming vectors and $\mathbf{H}[m] \in \mathbb{C}^{N_R \times N_T}$ channel matrix at time m .

- Beam Tracking:

$$\min_{g, \theta_T, \theta_R} \mathbb{E} \|\mathbf{H}[m] - \hat{\mathbf{H}}[m]\|_F$$

- Baseline:

- Exhaustive Search
- Complexity: $\mathcal{O}(N^2)$



System Model: Power Allocation

- Power Allocation for Beam Tracking (full CSI):

$$C = \sum_{i=1}^{n_{min}} \log\left(1 + \frac{P_i |h_i|^2}{N_0}\right) \text{ subject to } P = \sum P_i^*$$

- Optimum allocation: $P_i^* = \left(\frac{1}{\mu} - \frac{N_0}{|h_i|^2}\right)^+$
- Let $\mathbf{H} = \mathbf{U} \Sigma \mathbf{V}^* = \sum_{i=1}^{n_{min}} \sigma_i \mathbf{u}_i \mathbf{v}_i^*$; $n_{min} = \min(NT, NR)$

- At high SINR:

$$C = \sum_{i=1}^K \log\left(1 + \frac{1}{K} \frac{P_i |\sigma_i|^2}{N_0}\right) \approx K \times \log(\text{SNR}) + \sum_{i=1}^K \log\left(\frac{|\sigma_i|^2}{K}\right)$$

where $K = \text{rank}(\mathbf{H})$

Equal Power Allocation

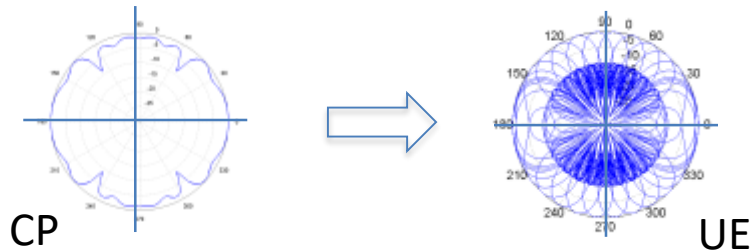
- At low SINR:

$$C = \sum_{i=1}^K \log\left(1 + \frac{1}{K} \frac{P_i |\sigma_i|^2}{N_0}\right) \approx \text{SNR} \left(\max_i \sigma_i^2\right) \log_2 e$$

Strongest Eigenmode

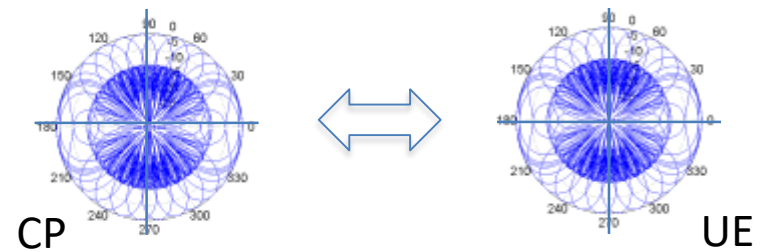
Beam Search: M+N

Stage 1:

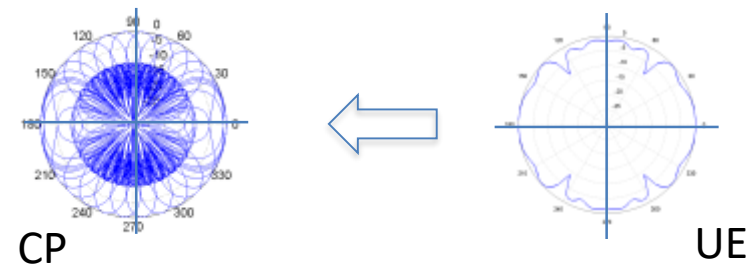


K out of N_{UE} best SINR beams

Stage 2:



$K_{CP} \times K_{UE}$ search



K out of M_{CP} best SINR beams

$$SINR_k = 10 \log \left(\frac{|\mathbf{w}_{R,k}^* \mathbf{H} \mathbf{w}_{T,k}|^2 P_K}{\sum_{i \neq k} |\mathbf{w}_{R,i}^* \mathbf{H} \mathbf{w}_{T,i}|^2 + N_0} \right)$$

- Number of Messages Tx'd:

$$M_{CP} + N_{UE} + K_{CP} \times K_{UE}$$

where $M_{CP} = 15, N_{UE} = 15, K_{CP} = 4, K_{UE} = 4$

- Search Time:

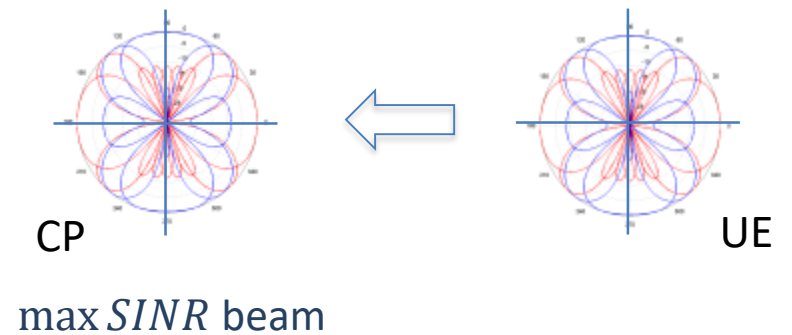
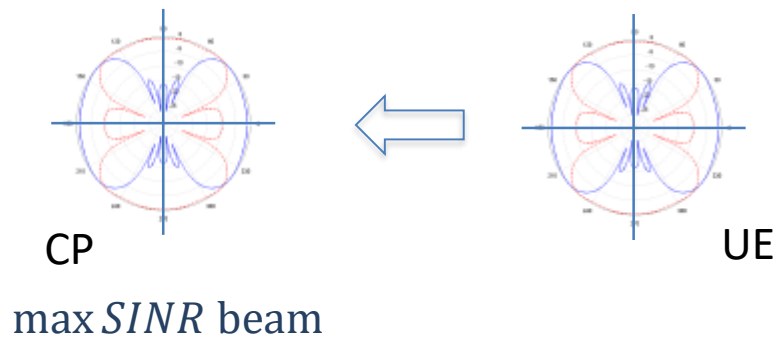
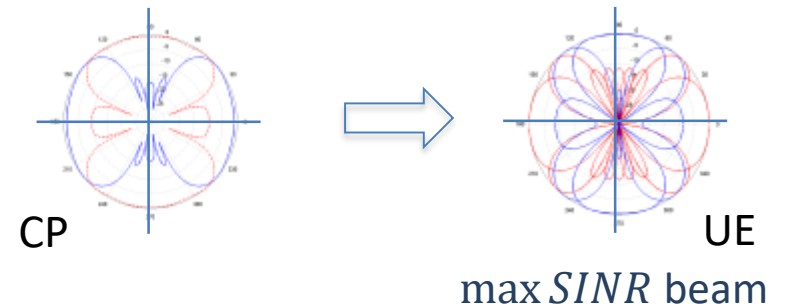
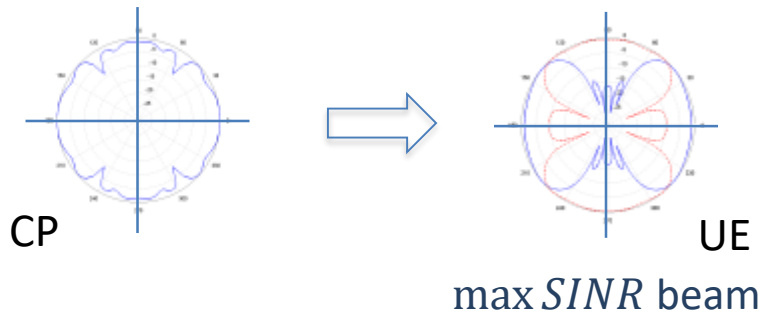
$$G_{128} M_{CP} + G_{128} N_{UE} + G_{64} (K_{CP} \times K_{UE}) \approx \mathbf{0.35 \mu s}$$

where G_N is a Golay sequence of length N .

Beam Search: Hierarchical

Omni Stage:

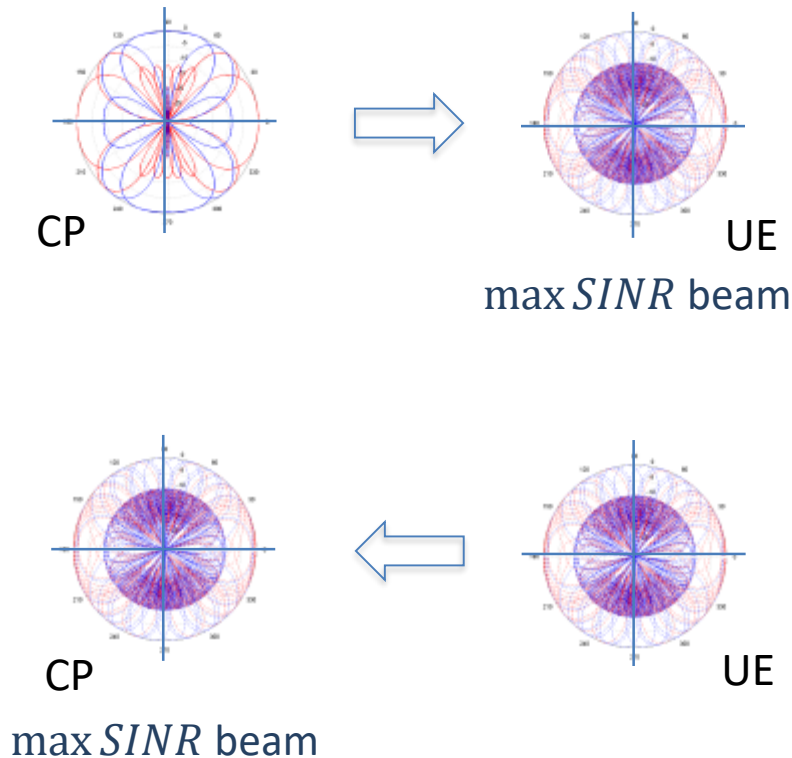
Sector Stage:



$$SINR_k = 10 \log \left(\frac{|\mathbf{w}_{R,k}^* \mathbf{H} \mathbf{w}_{T,k}|^2 P_K}{\sum_{i \neq k} |\mathbf{w}_{R,i}^* \mathbf{H} \mathbf{w}_{T,i}|^2 + N_0} \right)$$

Beam Search: Hierarchical

Beam Stage:



- Number of Messages Tx'd:

$$N_{zc} + N_{omni} + N_{sector} + N_{beam}$$

where $N_{zc} = 1, N_{omni} = 2, N_{sector} = 2, N_{beam} = 8$

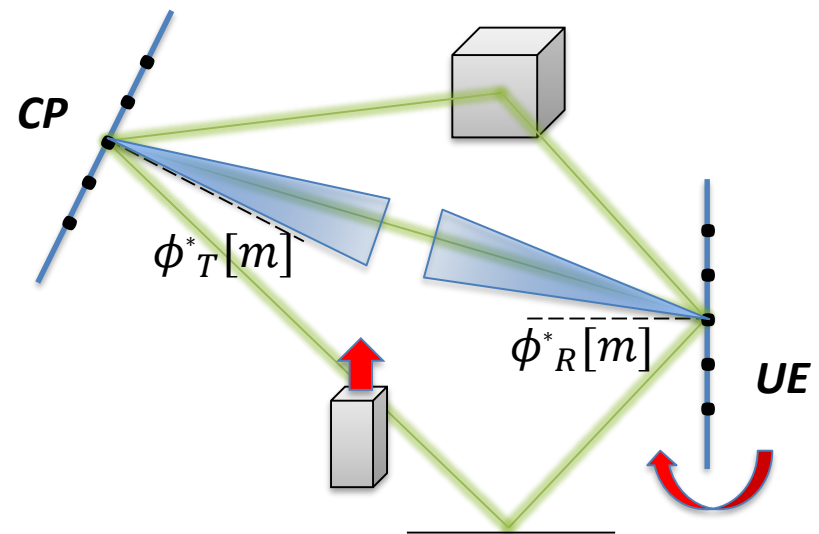
- Search Time:

$$G_{128}(2N_{zc}) + G_{128}(2N_{omni}) + G_{64}(N_{sector} + N_{beam}) + G_{32}(N_{beam}) \approx \mathbf{0.12 \mu s}$$

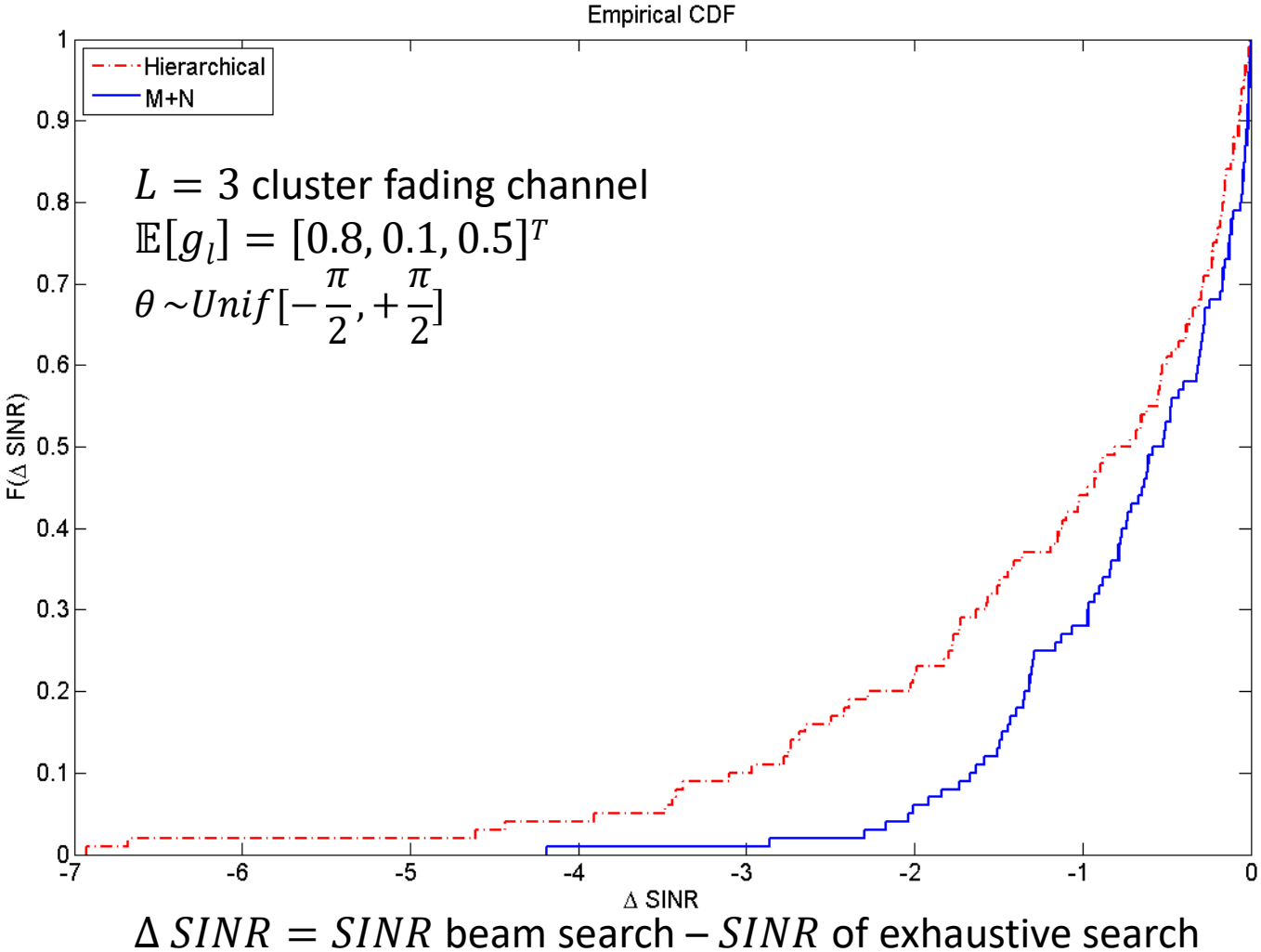
where G_N is a Golay sequence of length N .

Beam Tracking

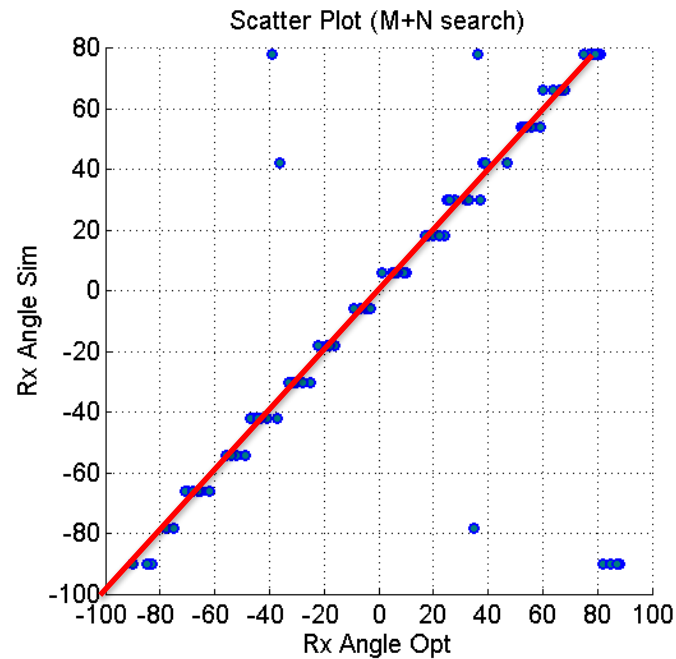
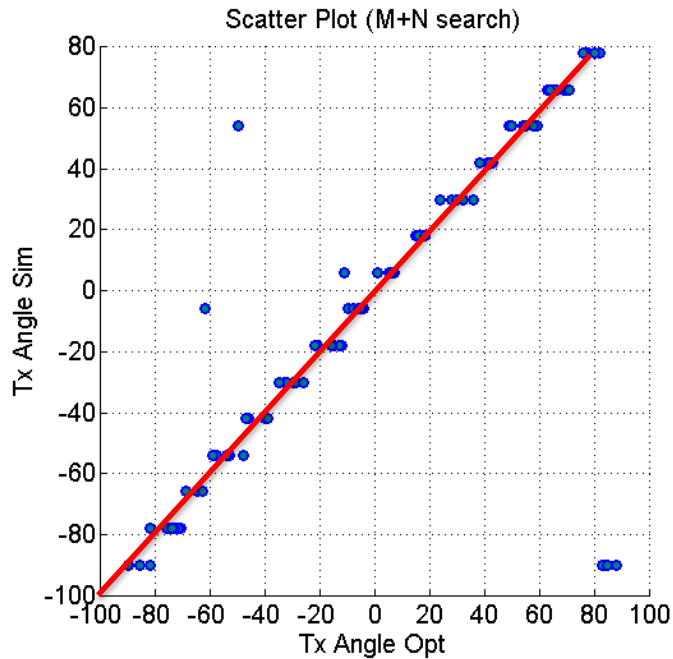
- Beam Switching:
 - Switch to 1 of K beams in a group of high SINR beams
 - Track SINR for the K beam group
- Additional Techniques:
 - Kalman Filter
 - Compressive Tracking [5]
 - Finite Rate of Innovation [6]



Simulation Results: Empirical CDF

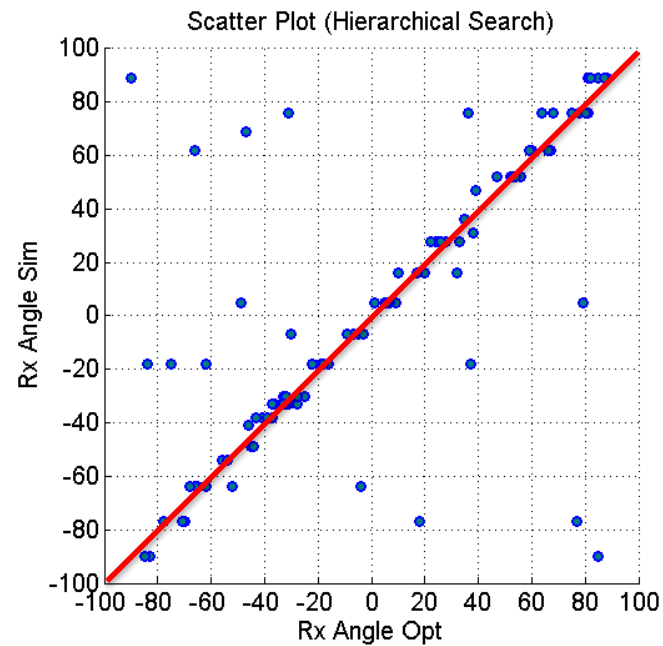
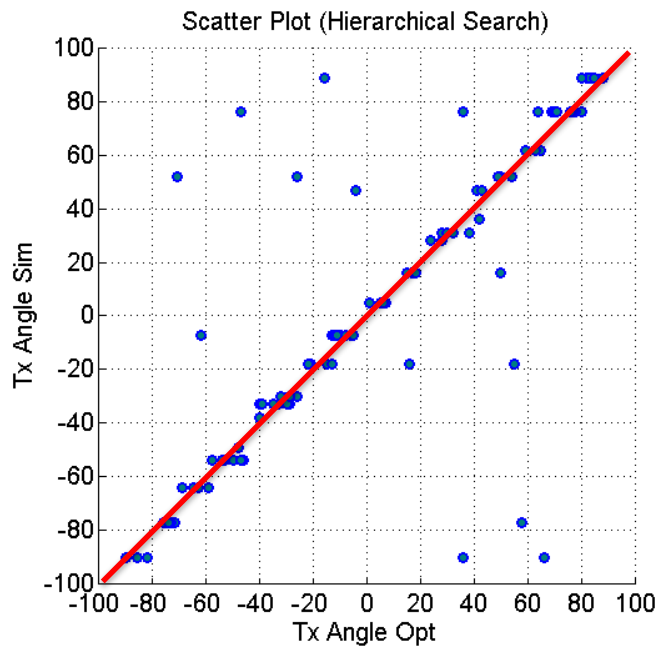


Simulation Results: Scatter Plot



If points are on the diagonal : ***M+N search*** found the correct beam angle

Simulation Results: Scatter Plot



If points are on the diagonal : **hierarchical search** found the correct beam angle

Discussion: Beam Search

M+N Beam Search

- $\mathbb{P}\{\Delta SINR < -3.01 \text{ dB}\} = 0.01$
- $\mathbb{P}\{\hat{\phi}_T \neq \phi_T\} = 0.07$
- $\mathbb{P}\{\hat{\phi}_R \neq \phi_R\} = 0.08$
- Simpler to configure
(fewer parameters)
- Finds K strongest paths during the last search stage for beam tracking
- $T_{mpn} = 0.35 \mu s$

Similarities:

- Overlapping beams reduce $\mathbb{P}_{\Delta SINR}$
- $\mathbb{P}_{\Delta SINR}$ can be improved by increasing the number of beams at the expense of increased complexity

Hierarchical Search

- $\mathbb{P}\{\Delta SINR < -3.01 \text{ dB}\} = 0.1$
- $\mathbb{P}\{\hat{\phi}_T \neq \phi_T\} = 0.13$
- $\mathbb{P}\{\hat{\phi}_R \neq \phi_R\} = 0.16$
- More involved: requires omni, sector and beam pattern design
- Finding K strongest paths requires re-running the full search
- Often finds 2nd strongest (reflected) path resulting in higher $\mathbb{P}_{\Delta SINR}$
- $T_{hierarchical} = 0.12 \mu s$

Discussion: Beam Tracking

Beam Tracking:

- Propagation may change very quickly due to motion, foliage blocks, hand/human body blockage (high reflectivity) [include diagram of cars and CPs from review slides or your own]
- Requires ability to quickly switch / track changing direction
- Parameter learning techniques may not be computationally fast enough.

Robustness:

- Environmental factors (e.g. wind, gust, rain) can disturb ϕ_T by $\pm 5^\circ$
- HPBW = -10; losses expected; formula for beam width
- Beam search and tracking can be controlled at lower carrier frequencies.

Discussion: mm-wave access

- CAP-MIMO: Increasing number of antennas requires, dedicated ADC/DAC chains –infeasible as N increases (esp. for UE) [11] (Beamforming by antenna weights/phase shifts (No separate RF chains)
- Noise-limited rather than interference limited
- Large bandwidth -> power limited regime (BPSK, QPSK constellations, good PAPR properties)
- High reflectivity of materials supports outdoor-outdoor communication
- Propagation may change very quickly due to motion, foliage blocks, hand/human body blockage (high reflectivity)

Future Work

- Multiple CP simulation of UE connected to multiple CPs (network level diversity) (slide 37 review)
 - Dense deployment of CPs, fast handover, diversity at the network level
- Multiple UE simulation:
 - UE to UE relaying for UEs without strong paths to any CP
 - Inter-CP load balancing due to limited backhaul
- Research and comparison of beam tracking techniques
- Multiplication of point sources by actual antenna responses
- Extension to large antenna arrays (CP) and realistic number of antennas (UE)
- Different mobility models / studies
- Large antenna arrays with coarse beamforming
- Understanding angular diversity (averaging over available angles / DoFs)

References

- [1] J. Brady, N. Behdad, A. Sayeed, “Beamspace MIMO for Millimeter-Wave Communications: System Architecture, Modeling, Analysis and Measurements,” *IEEE Transactions on Antennas and Propagation*, pp. 1-13, March 2013.
- [2] M. Akdeniz, Y. Liu, S. Rangan, E. Erkip, “Millimeter Wave Picocellular System Evaluation for Urban Deployments,” *CoRR*, pp. 1-9, April 2013.
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- [4] L. Du, et. al, “Review of User Parameter-Free Robust Adaptive Beamforming Algorithms,” *Digital Signal Processing*, 19(4), pp. 567-582, July 2009.
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- [10] C. Balanis, “Antenna Theory: Analysis and Design”, 3rd ed., ISBN 978-047-166-782-7, Apr 2005.
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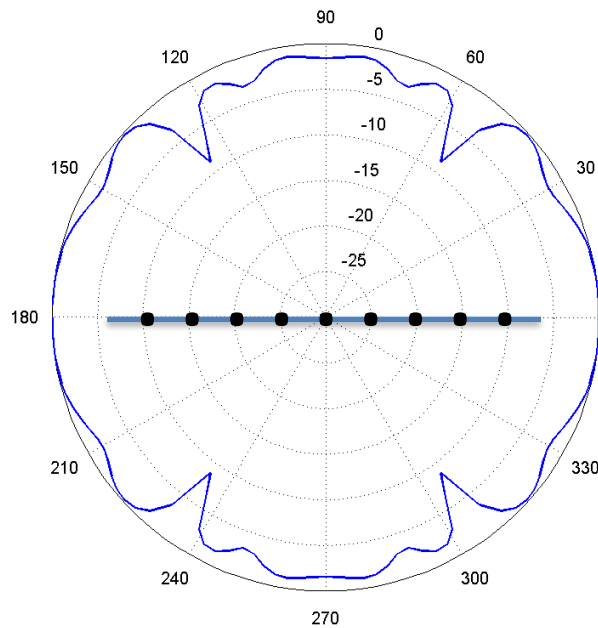
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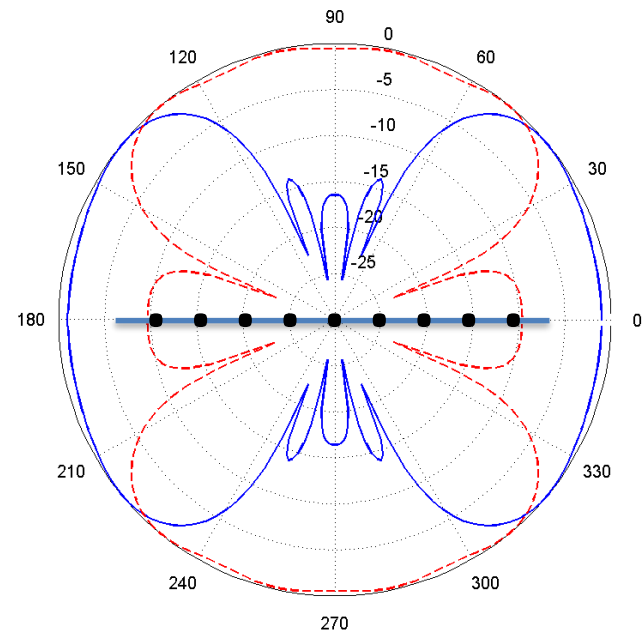
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Backup Slides: Beam Patterns

Zadoff-Chu



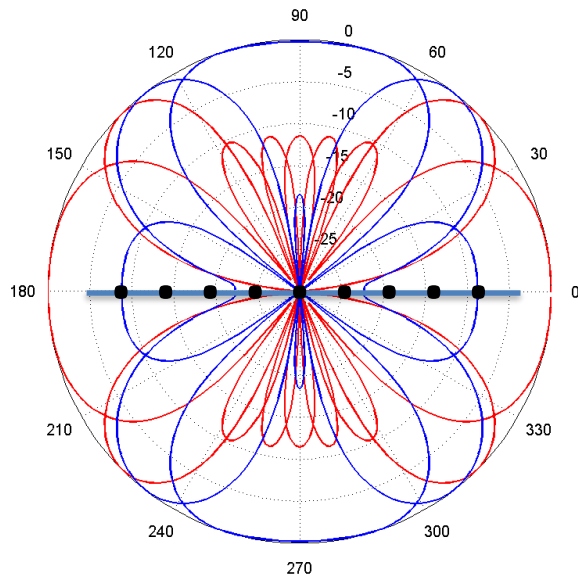
½ Omni



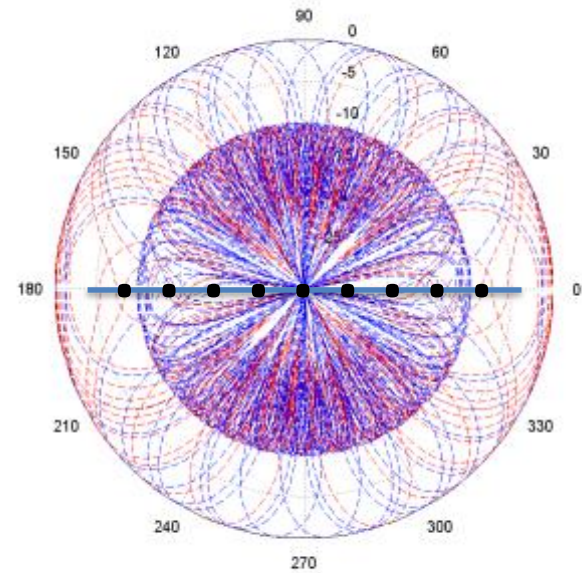
$$z_c[m + 1] = e^{-\frac{j\pi R m(m+1)}{N}}, \text{ for } m = 0, \dots, N - 1$$

Backup Slides: Beam Patterns

Sector



Beam



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